

9.2a) If  $\rho = 1$  then  $Z_1 = Z_2$ , so

$$Y_1 = X + Z_1$$

$$Y_2 = X + Z_1 = Y_1$$

So  $Y_2$  provides no additional information over  $Y_1$ .

$$\Rightarrow C = \frac{1}{2} \log\left(1 + \frac{P}{N}\right)$$

c) Here  $Z_2 = -Z_1$

$$\Rightarrow Y_1 = X + Z_1$$

$$Y_2 = X - Z_1$$

$$\Rightarrow \frac{Y_1 + Y_2}{2} = X,$$

a noise free observation of  $X$

$$\Rightarrow C = \infty$$

$$9.3 \quad C = \max_{P_X: EY^2 \leq P} I(X; Y)$$

$$\begin{aligned} I(X; Y) &= h(Y) - h(Y|X) \\ &\leq \frac{1}{2} \log 2\pi e P - h(Y|X) \\ &= \frac{1}{2} \log 2\pi e P - h(Z) \\ &= \frac{1}{2} \log 2\pi e P - \frac{1}{2} \log(2\pi e \sigma^2) \\ &= \frac{1}{2} \log P/\sigma^2 \end{aligned}$$

We have equality when  $Y \sim N(0, P)$   
 which happens when  $X \sim N(0, P - \sigma^2)$

9.5  $Y = XV + Z$

$$\begin{aligned} I(X; Y) &\leq I(X; Y, V) \\ &= \underbrace{I(X; V)}_{=0} + I(X; Y|V) \\ &= I(X; Y|V) \end{aligned}$$

where  $I(X; V) = 0$  since  $X$  &  $V$  are independent

9.6 Recall that

$$P_1 = E X_1^2$$

$$P_2 = E X_2^2$$

$$\text{and } P_j = [v - \sigma_j^2]^+ \quad j=1,2$$

We are at the border of assigning power to channel 1 when  $v = \sigma_1^2$

$$\Rightarrow P_2 = \sigma_1^2 - \sigma_2^2 \quad \text{and} \quad P_1 = \sigma_1^2 - \sigma_1^2 = 0$$

$$\Rightarrow E X_1^2 + E X_2^2 = \sigma_1^2 - \sigma_2^2$$

9.7

$$\begin{aligned} \text{a) } Y &= (X + Z_1) + (X + Z_2) \\ &= 2X + Z_1 + Z_2 \end{aligned}$$

$$\text{Signal Power: } E (2X)^2 = 4\rho^2$$

$$\text{Noise Power: } E (Z_1 + Z_2)^2 = \sigma^2 + 2\rho\sigma^2 + \sigma^2$$

$$\Rightarrow C = \frac{1}{2} \log \left( 1 + \frac{4\rho}{2\sigma^2 + 2\rho\sigma^2} \right)$$

$$= \frac{1}{2} \log \left( 1 + \frac{2\rho}{\sigma^2(1+\rho)} \right)$$

$$\text{b) } \rho = 0 \Rightarrow C = \frac{1}{2} \log \left( 1 + \frac{2\rho}{\sigma^2} \right)$$

$$\rho = 1 \Rightarrow C = \frac{1}{2} \log \left( 1 + \frac{\rho}{\sigma^2} \right)$$

$$\rho = -1 \Rightarrow C = \infty$$

$$\begin{aligned}
 \underline{9.12} \quad hR = H(w) &= \mathcal{I}(w; \hat{w}) + H(w | \hat{w}) \\
 &\leq \mathcal{I}(w; \hat{w}) + n\epsilon_n \\
 &\leq \mathcal{I}(x_1^n; \gamma_1^n) + n\epsilon_n \\
 &= h(\gamma_1^n) - h(\gamma_1^n | x_1^n) + n\epsilon_n \\
 &\leq \sum_{\ell=1}^n h(\gamma_\ell) - h(z_\ell) + n\epsilon_n \\
 &= \sum_{\ell=1}^n \mathcal{I}(x_\ell; \gamma_\ell) + n\epsilon_n
 \end{aligned}$$

Now,  $\frac{1}{n} \sum_{\ell=1}^n \gamma_\ell^2(w) \leq \rho$

$$\Rightarrow \frac{1}{n} \sum_{\ell=1}^n E[X_\ell^2] \leq \rho$$

$$\Rightarrow E[X_\ell^2] \leq n\rho \quad \text{for } \ell = 1, \dots, n$$

$$\Rightarrow E\left[\left(\frac{X_\ell}{\ell}\right)^2\right] \leq \frac{n\rho}{\ell^2}$$

$$\Rightarrow hR \leq \sum_{\ell=1}^n \frac{1}{2} \log\left(1 + \frac{n\rho}{\ell^2 N}\right) + n\epsilon_n$$

$$R \leq \frac{1}{2} \cdot \frac{1}{n} \sum_{\ell=1}^n \log\left(1 + \frac{n\rho}{\ell^2 N}\right) + \epsilon_n$$

Now,  $\tilde{\epsilon}_n \rightarrow 0$  as  $n \rightarrow \infty$ , and

$$\begin{aligned} & \frac{1}{n} \sum_{l=1}^n \log\left(1 + \frac{nP}{e^{2N}}\right) \\ & \leq \frac{1}{n} \sum_{l=1}^{\infty} \log\left(1 + \frac{nP}{e^{2N}}\right) \\ & \rightarrow 0 \quad \text{as } n \rightarrow \infty. \end{aligned}$$

This is because

$$\begin{aligned} & \lim_{n \rightarrow \infty} \frac{\sum_{l=1}^{\infty} \log\left(1 + \frac{nP}{e^{2N}}\right)}{n} \\ & = \lim_{n \rightarrow \infty} \left[ \frac{\sum_{l=1}^{\infty} 1}{1 + \frac{nP}{e^{2N}}} \cdot \frac{P}{e^{2N}} \right] \\ & \quad \text{---} \\ & \quad \quad \quad 1 \end{aligned}$$

by L'Hôpital's rule and this limit is 0.

$$\Sigma_0 \quad R \leq 0$$

9.14  $C \geq I(X; Y)$  for any  $X \sim P_X$ .

Let  $X \sim N(0, P)$ . Then  $h(Y)$  is finite.

$$\text{Now } f_Z(z) = \frac{1}{10} \delta(z) + \frac{9}{10} g_{Z^*}(z)$$

where  $g_{Z^*}(z)$  is pdf of  $Z^* \sim N(0, V)$

$$\begin{aligned} \text{Then } h(Z) &= - \int f_Z(z) \log f_Z(z) dz \\ &\leq - \int \frac{1}{10} \delta(z) \log \left[ \frac{1}{10} \delta(z) \right] dz \\ &= -\infty \end{aligned}$$

So  $C \geq h(Y) - [-\infty] = \infty$ .

The following strategy works:

Let  $B_1, B_2, B_3, \dots$  be the infinite sequence of bits to transmit in a finite time  $n$ .

Construct the random number  $T = \sum_{l=1}^{\infty} B_l / 2^l$

$$\text{Pick } x = \frac{\sqrt{P} \cdot T}{E[T^2]}$$

$$\text{Then } E x^2 = P$$

Also, from  $T$  (or equivalently  $x$ ), we can recover the infinite # of bits error free.

To send  $x$  ~~error~~ error free with high probability, choose the inputs  $x_1 = x_2 = \dots = x_n = x$ .

The receiver looks for 2  $Y$ 's, such that

$$Y_i = Y_j \text{ for } i \neq j.$$

If this happens it is because  $Y_i = Y_j = x$ , and the receiver knows  $x$  now.

The probability of this not happening decays exponentially to 0 as  $n \rightarrow \infty$ .